

This lecture:

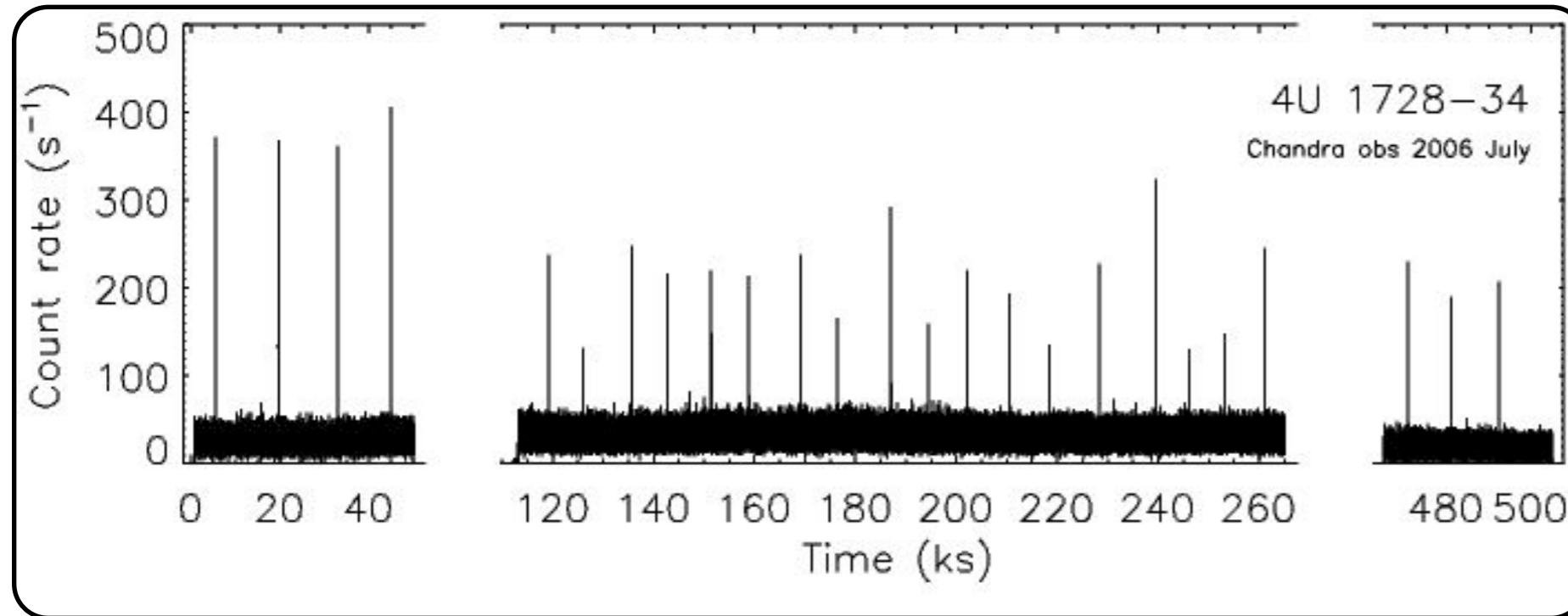
Total overview of a measurement procedure

Dealing with noise, e.g. optimal filtering

Signal-to-noise ratios

Is radiation indeed intrinsically stochastic?

One bin/measurement: average in time domain over box



Type I X-ray bursts

GALLOWAY PRIVATE COMMUNICATION

1 MEASUREMENT OF $x(t)$ IN A TIME $T \Rightarrow$ WINDOWING AND
AVERAGING OVER TIME ΔT

WINDOWING: $y(t) = \Pi\left(\frac{t}{T}\right)x(t)$

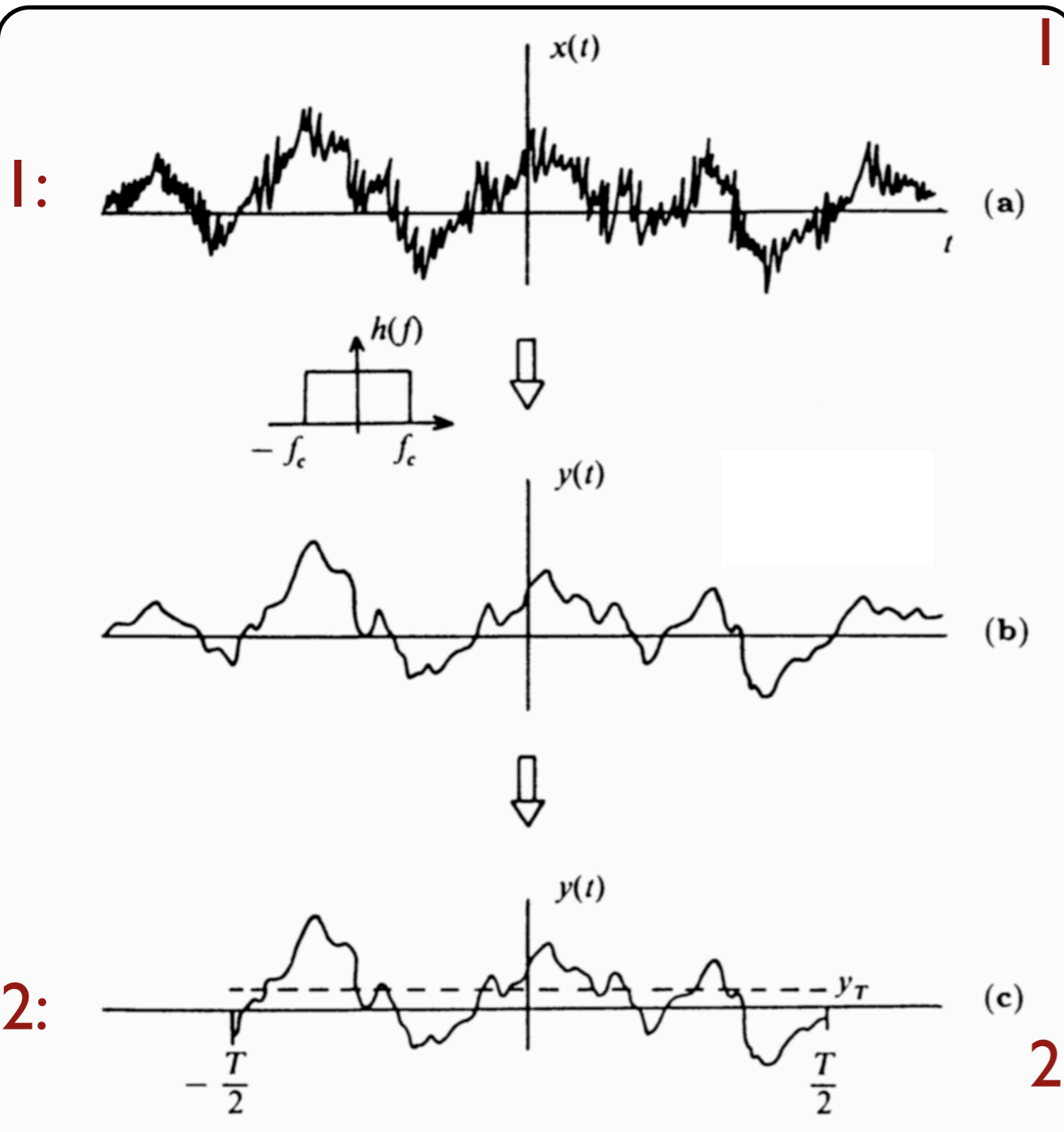
AVERAGING:

$$z(t) \equiv y_{\Delta T}(t) = \frac{1}{\Delta T} \int_{t-\Delta T/2}^{t+\Delta T/2} y(t') dt' = \frac{1}{\Delta T} \int_{-\infty}^{\infty} \Pi\left(\frac{t-t'}{\Delta T}\right) y(t') dt'$$

This is a
convolution!



= LOW-PASS FILTER, REMEMBER NYQUIST THEOREM



1: RESPONSE → REMOVE HIGH FREQUENCIES DUE TO FINITE RESPONSE OF INSTRUMENT AND DETECTOR

Remember convolution goes from $-\infty$ to ∞

2: 1 SAMPLE → AVERAGING →

Average in time domain over length T , i.e. convolve in time domain with box function

Note figure used here different from Lecture notes p25!

$$z(t) = \frac{1}{\Delta T} \Pi\left(\frac{t}{\Delta T}\right) * y(t)$$
$$z(t) = \frac{1}{\Delta T} \Pi\left(\frac{t}{\Delta T}\right) * \Pi\left(\frac{t}{T}\right) x(t)$$

What is the combined effect of finite sample and finite bin size?

Consider example on pages 25-27 of Lecture notes

$$\sigma_{x_T}^2 = \sigma^2 / N \text{ for } T \gg \tau_0$$

ESTIMATING σ IF THE PROBABILITY DENSITY
FUNCTION OF THE S.P. IS KNOWN



PROPAGATION OF ERRORS **CHAPTER 5.2**

ESTIMATING σ IF THE PROBABILITY DENSITY
FUNCTION OF THE S.P. IS NOT KNOWN



E.G.

BOOTSTRAP METHOD

JACKKNIFE METHOD

MORE ON THIS LATER

SIGNAL DETECTION INVOLVES:

 LIMITED TIME INTERVAL → WINDOWING

 NOT CONTINUOUS → SAMPLING

 SAMPLES NOT INSTANTANEOUS → AVERAGING

 DEALING WITH NOISE → FILTERING

 RESPONSE OF THE DETECTION SYSTEM



THUS THE DETECTED SIGNAL WILL ONLY
APPROXIMATE THE SOURCE SIGNAL

Recap

SAMPLING: HIGH FREQUENCIES ARE FILTERED OUT

WINDOW: LOW FREQUENCIES ARE FILTERED OUT



LEADS TO BAND LIMITED DATA

→ TIME

MEASURE A PROCESS $x(t)$ OVER INTERVAL T ASSUMED
ZERO OUTSIDE T

$$\equiv y(t) = \Pi\left(\frac{t}{T}\right)x(t)$$

$$Y(f) = X(f) * T \text{sinc}(Tf)$$

ALL INFORMATION ABOUT FREQUENCIES $< 1/T$ IS LOST!

Recap

SAMPLING: HIGH FREQUENCIES ARE FILTERED OUT

WINDOW: LOW FREQUENCIES ARE FILTERED OUT



LEADS TO BAND LIMITED DATA

→ FREQUENCY

$$Y(f) = X(f)H(f)$$

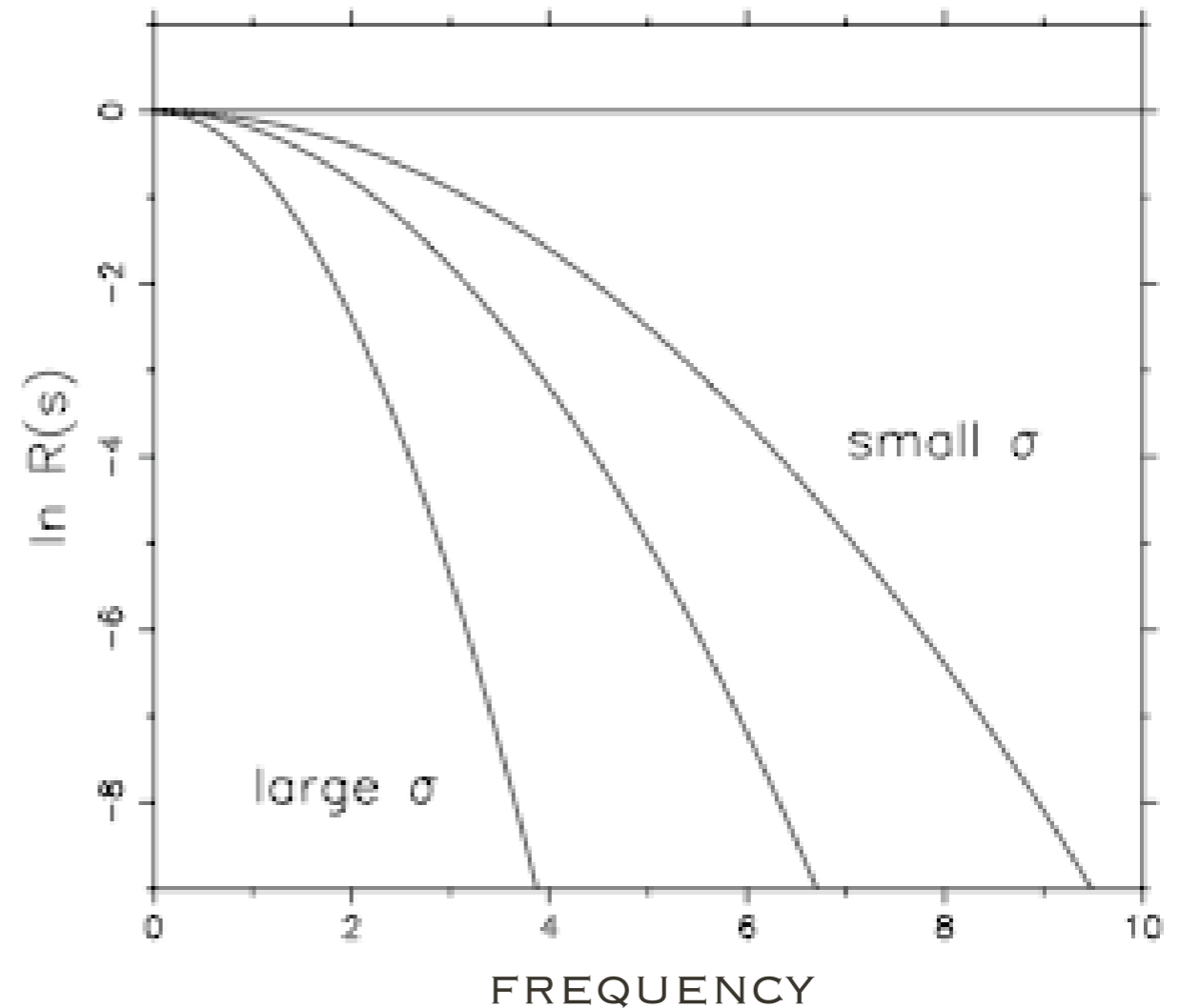
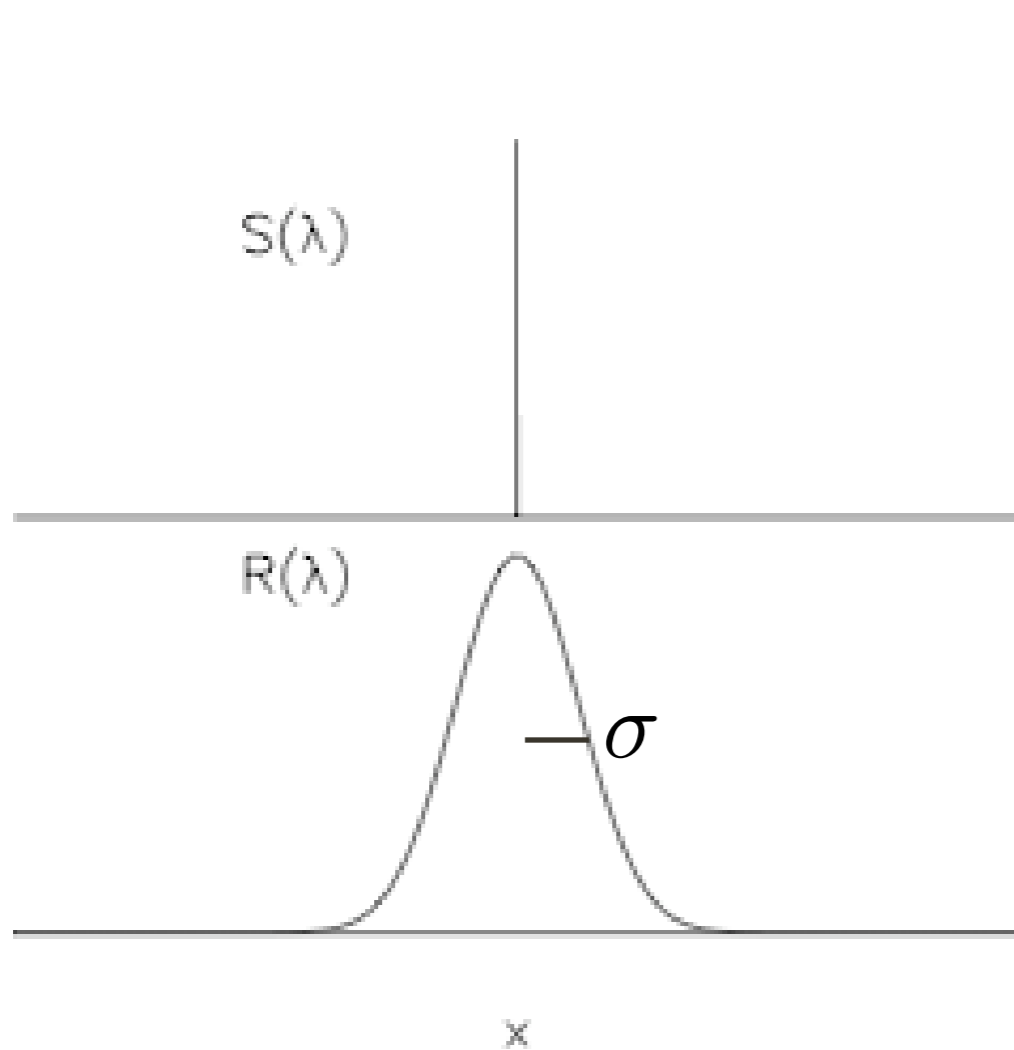
$$y(t) = \int_{-\infty}^{\infty} x(t - \theta)h(\theta)d\theta$$

$$y(t) = x(t) * h(t)$$

FILTERING OF PROCESS X WITH FILTER H

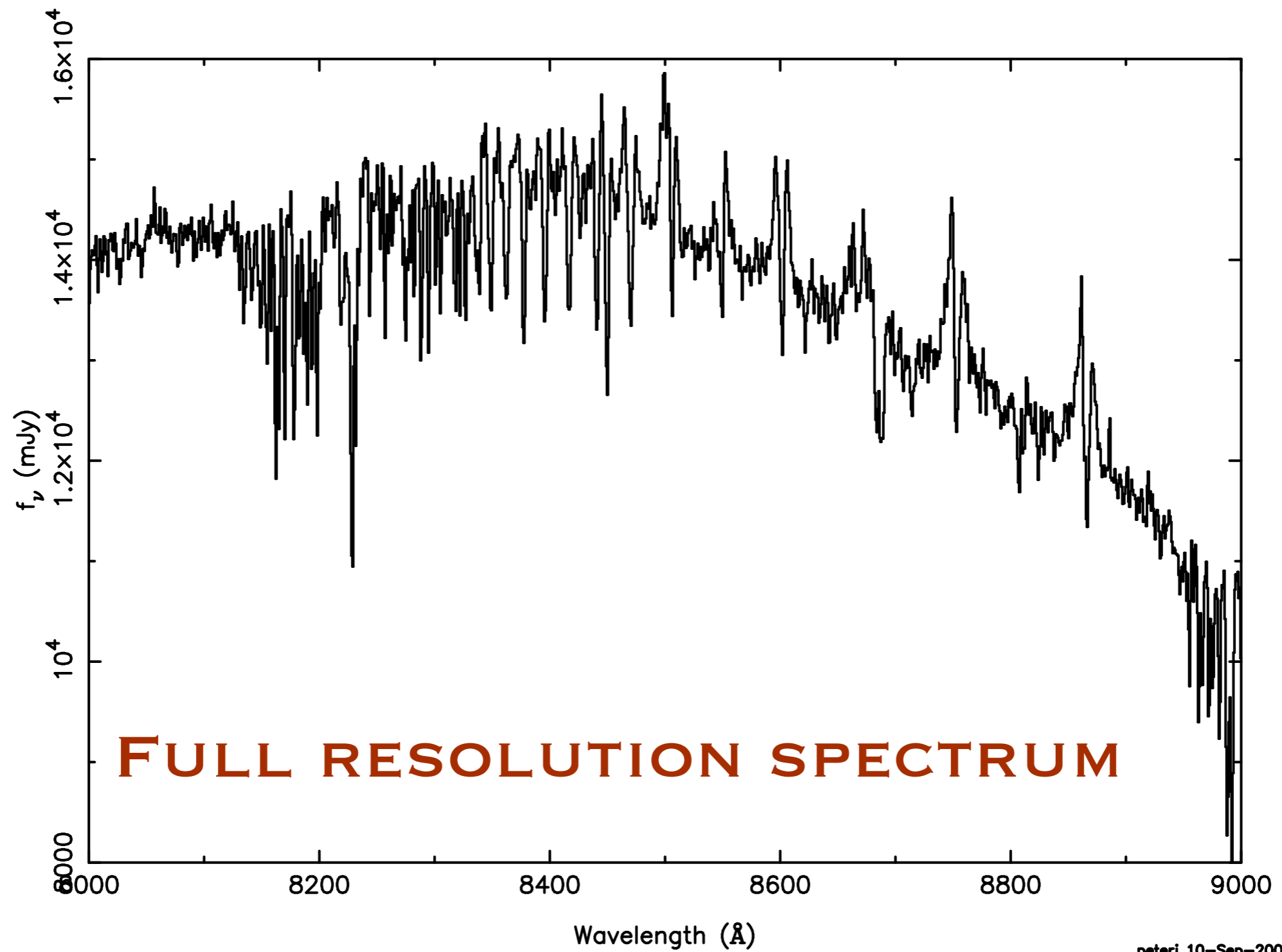
Recap: detector/instrument system response

GAUSSIAN RESPONSE FUNCTION

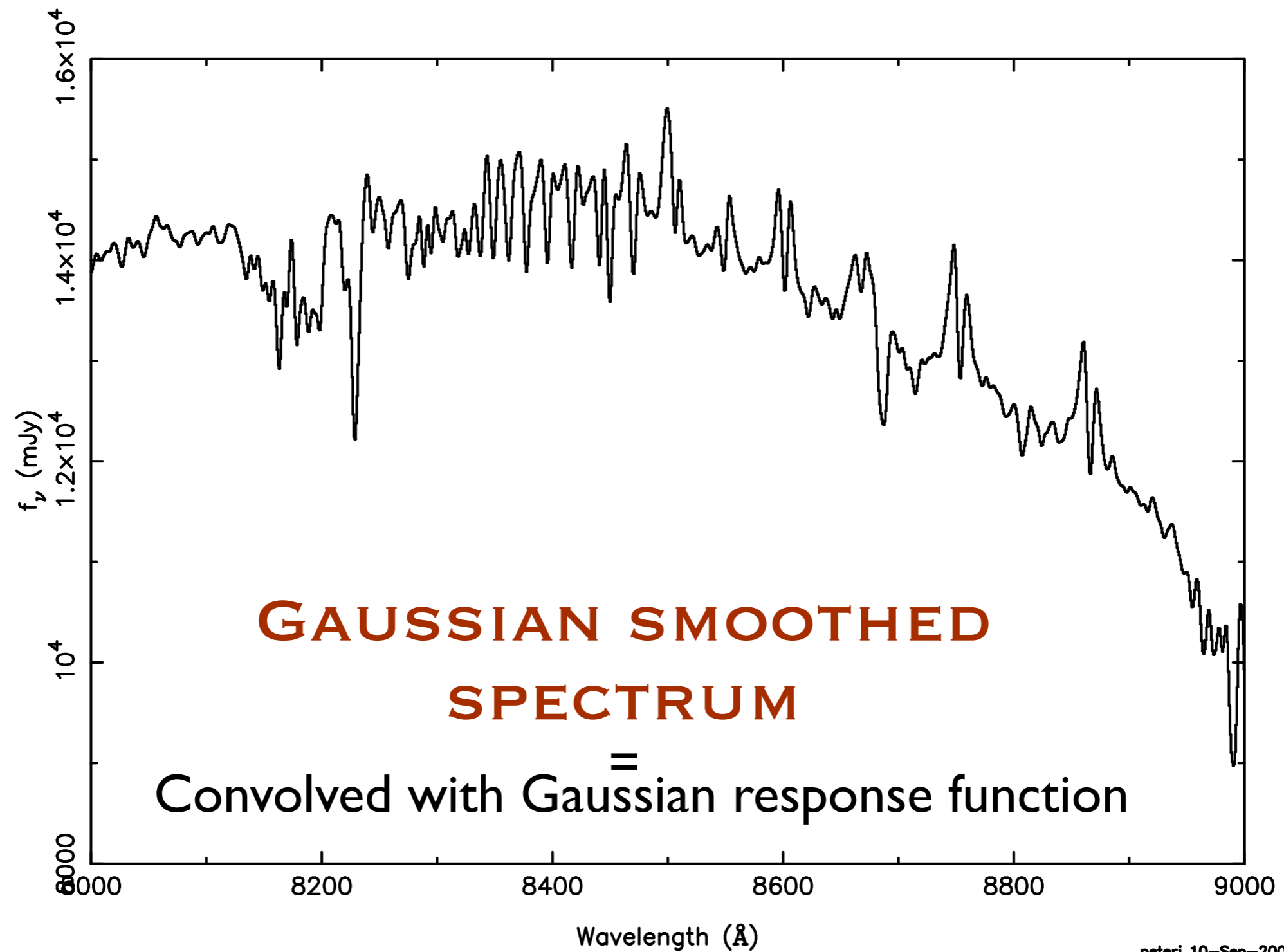


$$R(\lambda) = \frac{1}{\sqrt{(2\pi)\sigma}} \exp - \left(\frac{\lambda^2}{2\sigma^2} \right)$$

GAUSSIAN RESPONSE FUNCTION



GAUSSIAN RESPONSE FUNCTION



Mazur: Peer Instruction

Identify in a real astrophysical
measurement

all the steps discussed above

Response instrument/detector

Finite duration measurement

Sampling

Average of N samples

Noise contributions?

RECAP

DATA SAMPLING: IDEAL CASE NYQUIST CRITERIUM IS FULFILLED → SAMPLING DOES NOT LEAD TO LOSS OF INFORMATION RV completely described by the samples

CONDITIONS either:

→ BAND-LIMITED RESPONSE OF THE DETECTOR REMOVES HIGHEST NOISE POWERS AND THE SAMPLING IS FAST ENOUGH TO COVER THE BAND LIMIT OF THE DETECTOR or:

→ SIGNAL IS BAND-LIMITED and

$$\nu_{\text{sampling}} > \nu_{\text{max,detector}} > \nu_{\text{max,signal}}$$

however, noise is often white and thus contains high frequencies!

NOISE REMOVAL BY OPTIMAL FILTERING

Num. Res 13.3

$$c(t) = s(t) + n(t)$$

$s(t)$ is the smeared signal i.e. true * response

$$s(t) = u(t) * r(t)$$

DESIGN AN OPTIMAL FILTER $\phi(t)$ THAT
PRODUCES A SIGNAL $\tilde{u}(t)$ AS CLOSE AS
POSSIBLE TO $u(t)$ or $U(f)$

$$\tilde{U}(f) = \frac{C(f)\phi(f)}{R(f)}$$

CLOSE IN LEAST SQUARE SENSE

$$\int_{-\infty}^{\infty} |\tilde{U}(f) - U(f)|^2 df \text{ IS MINIMISED}$$

NOISE REMOVAL BY OPTIMAL FILTERING

$$\int_{-\infty}^{\infty} \left| \frac{[S(f) + N(f)]\phi(f)}{R(f)} - \frac{S(f)}{R(f)} \right|^2 df = 0$$

$\int S(f)N(f)df$ TERMS ARE ZERO SINCE NOISE
AND SIGNAL ARE UNCORRELATED

$$\int_{-\infty}^{\infty} |R(f)|^{-2} \underbrace{\{|S(f)|^2|1 - \phi(f)|^2 + |N(f)|^2|\phi(f)|^2\}}_{\Theta} df = 0$$

Θ MINIMISED WITH RESPECT TO ϕ

NOISE REMOVAL BY OPTIMAL FILTERING

$$\frac{d\theta}{d\phi} = 0$$

$$-2S^2(1 - \phi) + 2N^2\phi = 0$$

OPTIMAL FILTER

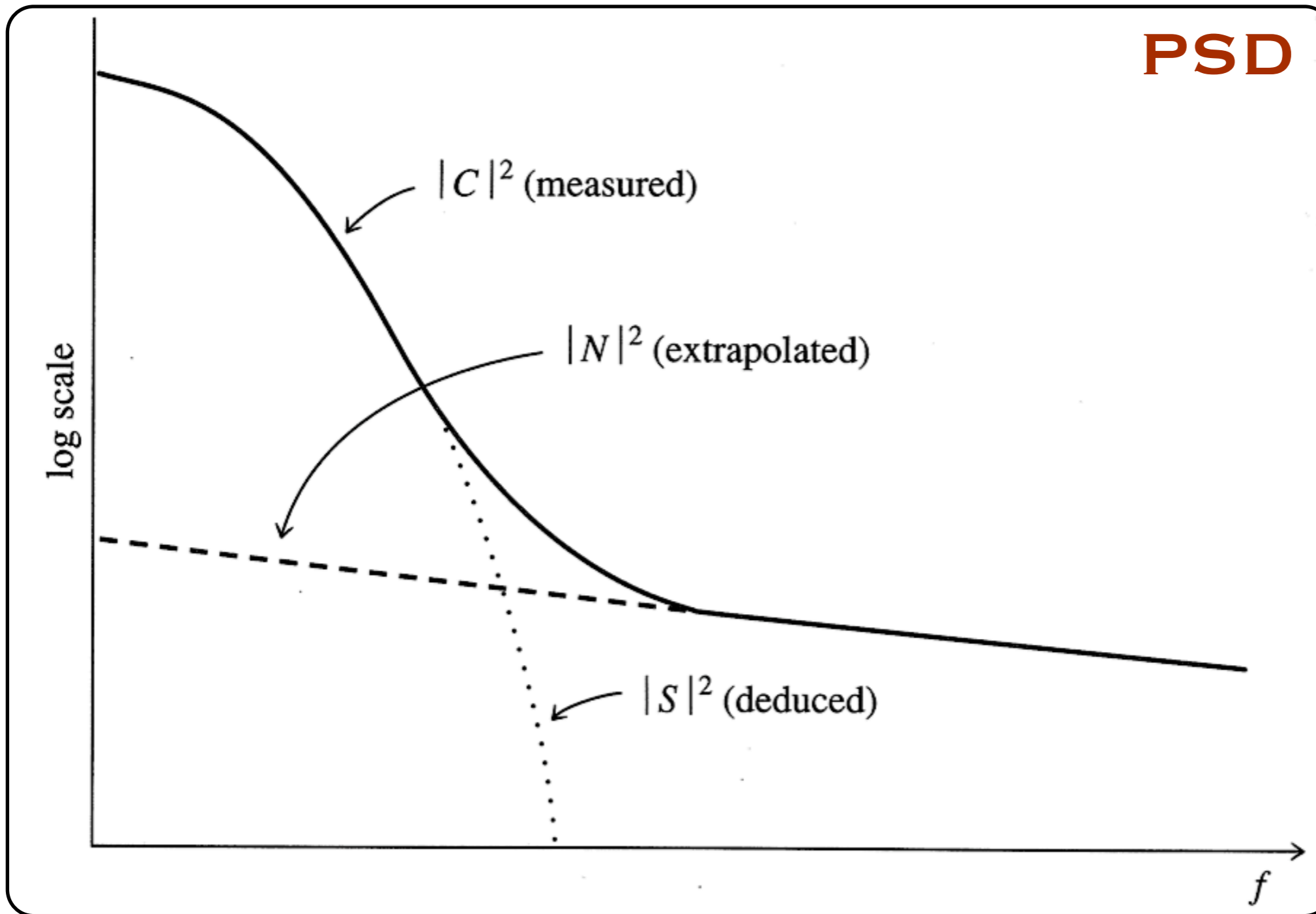
$$\phi = \frac{S^2}{S^2 + N^2}$$

DOES NOT CONTAIN TRUE SIGNAL DIRECTLY!
 $U(f)$

$$|S(f)|^2 + |N(f)|^2 = PDS(f) = |C(f)|^2$$

NOISE REMOVAL BY OPTIMAL FILTERING

PSD



SOME APPLICATIONS OF FILTERING

ON OPTIMAL DETECTION OF POINT SOURCES IN CMB MAPS

VIO ET AL, 2002, A&A, 391, 789

AN OPTIMAL FILTER FOR THE DETECTION OF GALAXY

CLUSTERS THROUGH WEAK LENSING

MATURI, ET AL. 2005, A&A, 442, 851

THE LARGEST SCALE PERTURBATIONS: A WINDOW ON THE

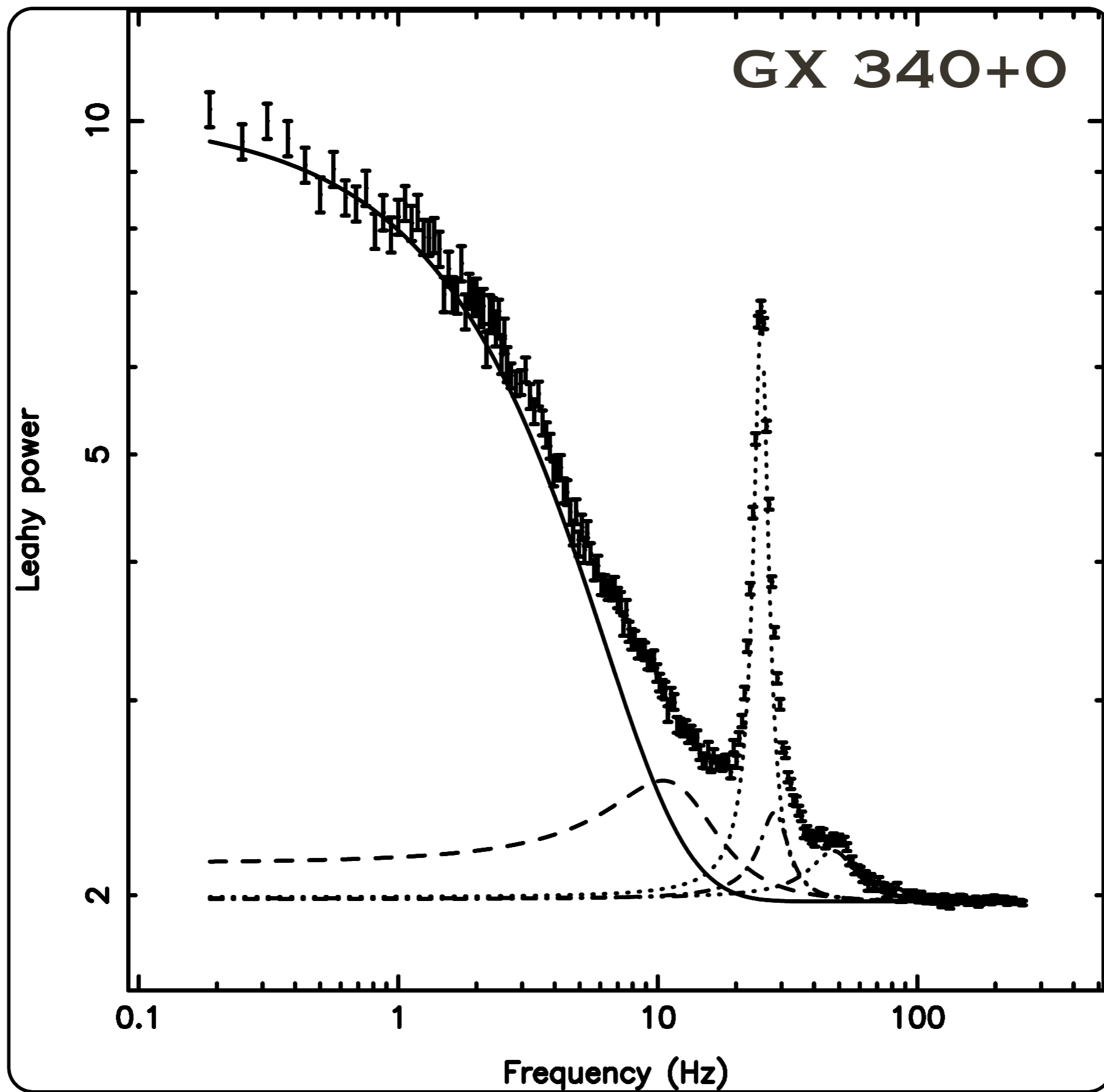
PHYSICS OF THE BEGINNING

WANDELT, NEW ASTRONOMY REVIEW, 2006, 11, 900

Application to timing experiments in Low-mass X-ray binaries

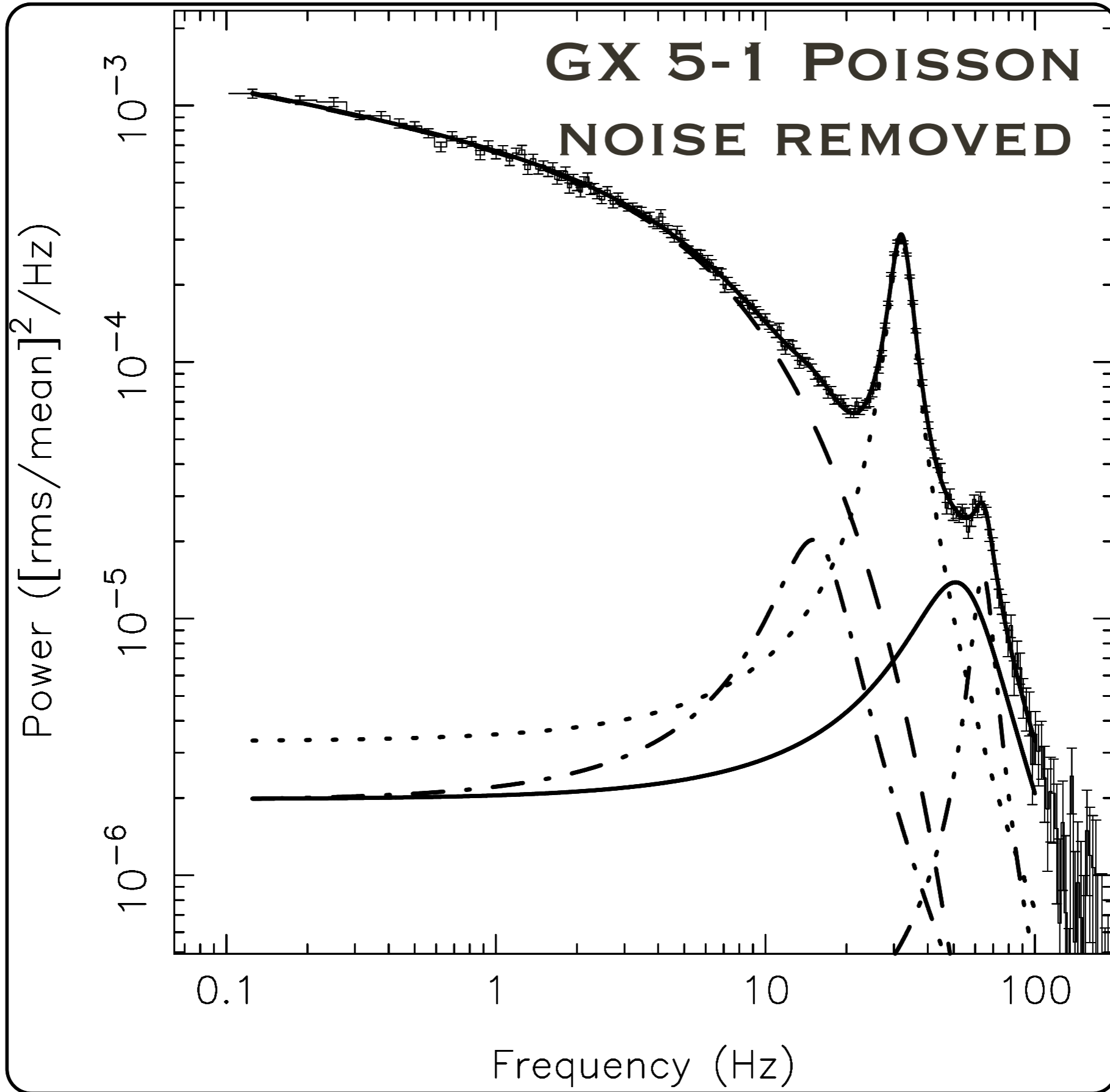


C.F. X-RAY TIMING EXPERIMENTS

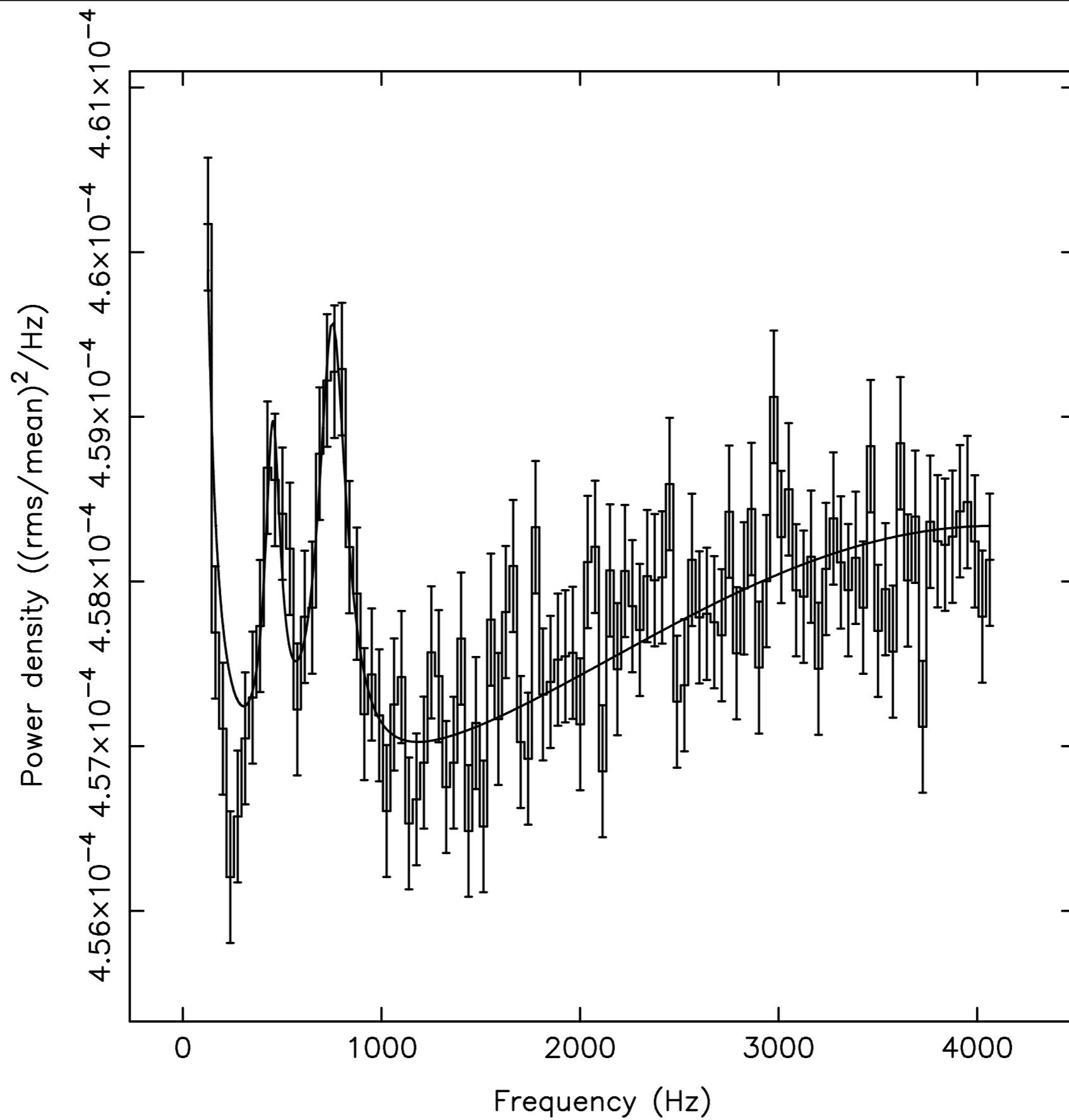


X-RAY TIMING EXPERIMENTS

**GX 5-1 POISSON
NOISE REMOVED**



X-RAY TIMING EXPERIMENTS



RECAP FILTERING:

ONE CAN DESIGN AN OPTIMAL FILTER SUCH THAT THE FILTERED MEASURED DATA-SET IS AS CLOSE AS POSSIBLE (IN LEAST-SQUARE SENSE) TO THE UNCORRUPTED SIGNAL

The Signal-to-Noise Ratio & Limiting Sensitivity

1. The level of noise determines the detectability of a weak (source) signal
2. Various noise contributions from :
 - a) other sources in FoV (point-source / diffuse) → sky-noise
 - b) operational environment of instrument (radiation belts, scatter light from Earth, Sun, Moon, atmosphere etc.)
 - c) constituting parts of instrument e.g. noise from amplifiers in electronics, induced radio-activity → instrumental noise
3. Quality of an observation is determined by the signal-to-noise ratio **SNR**

$$\text{SNR} = \text{SNR}(T_{\text{exp}}, \Delta E, A_{\text{eff}}, B)$$

4. For photon energies \gtrsim eV (i.e. optical, UV, X-rays, γ -rays) quantum characterisation applicable: statistical treatment of the accumulated quanta (Poissonian statistics)

- Consider point source embedded in uniform sky background noise with photon flux n_s (ph/m²s) i.e. the source-region
- Sky background noise n_{bg} (ph/m²s sr)
- Instrumental noise n_{det} (ph/m²s)

Number of counts registered in source region within observation time T_{obs} :

$$N_1 = (n_s + n_{bg} \Delta\Omega) \bar{A}_{eff} + n_{det} A_{pix} T_{obs}$$

$\Delta\Omega$ - solid angle subtended by source region

\bar{A}_{eff} - effective area (averaged over bandwidth ΔE : $\int A_{eff}(E) dE / \Delta E$)

A_{pix} - pixel surface

- Similar expression holds for a region excluding the source: the background region

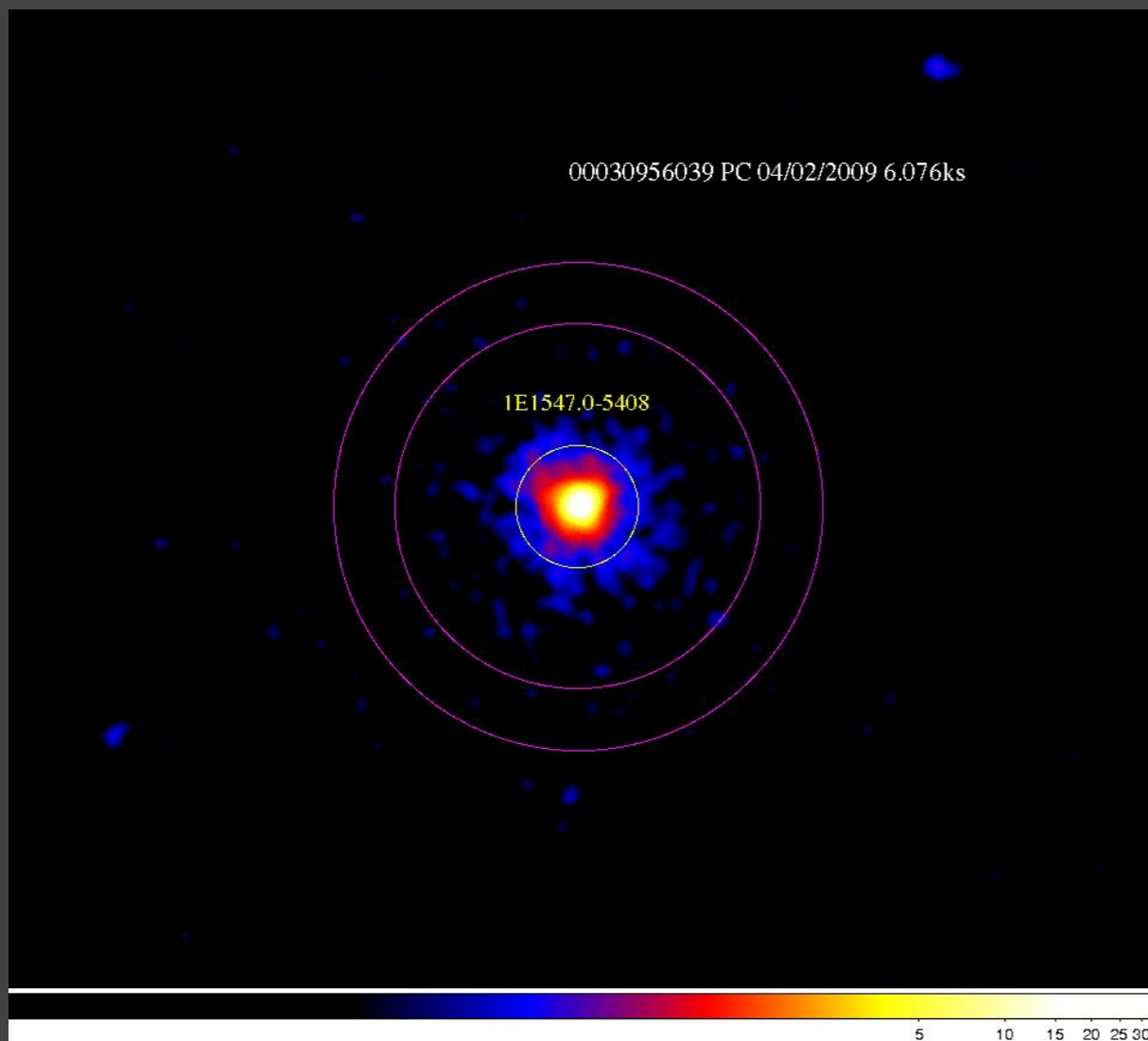
$$N_2 = (n_{bg} \Delta\Omega \bar{A}_{eff} + n_{det} A_{pix}) T_{obs}$$

collected during same integration time simultaneously

- Fluctuations in N_1 and N_2 are governed by Poissonian stat. : $\sqrt{N_1}$, $\sqrt{N_2}$

Swift-XRT observation of anomalous X-ray pulsar
1E1547.0-5408 during monitoring campaign after
its Jan. 2009 outburst (in PC-mode)

Example:



Inner (yellow) ring: 60'' source region
Purple annulus : 180''-240'' bakground region

Note: Background region 7x larger than source region

SNR can be defined by:

$$\text{SNR} = (N_1 - N_2) / \sqrt{(N_1 + N_2)}$$

Signal strength is expressed in terms of the fluctuation of the noise term(s)!

We speak of a 3σ measurement, if $\text{SNR} = k_{\min} = 3$ (k_{\min} is minimum confidence set by observer)

5σ measurement, if $\text{SNR} = k_{\min} = 5$

Limiting sensitivity

We consider now 2 limiting cases:

a) Source signal dominates the noise: $N_1 \gg N_2$

$$\text{SNR} \sim \sqrt{N_1} = \sqrt{(n_s \bar{A}_{\text{eff}} T_{\text{obs}})}$$

demanding a minimum SNR of k_{min} (typically 3 / 5 to speak of a detection) \rightarrow

$$n_{\text{smin}} = k_{\text{min}}^2 / (\bar{A}_{\text{eff}} T_{\text{obs}}) \quad \Rightarrow \text{the detection is } \underline{\text{signal-photon-noise}} \text{ limited}$$

b) Source signal drowned in the noise: $N_1 \approx N_2$

$$\text{SNR} = k_{\text{min}} \sim (n_s \bar{A}_{\text{eff}} T_{\text{obs}}) / \sqrt{(2N_2)} \Rightarrow$$

$$n_{\text{smin}} = k_{\text{min}} \sqrt{(2N_2)} / (\bar{A}_{\text{eff}} T_{\text{obs}}) \sim k_{\text{min}} \sqrt{(2\bar{A}_{\text{eff}} T_{\text{obs}} [n_{\text{bg}} \Delta\Omega + n_{\text{det}} \delta])} / \bar{A}_{\text{eff}} T_{\text{obs}} \propto (\bar{A}_{\text{eff}} T_{\text{obs}})^{-\frac{1}{2}}$$

$$(\delta = A_{\text{pix}} / \bar{A}_{\text{eff}})$$

n_{smin} improves only with the square root of the observation time and effective area...

If $n_{\text{det}} \delta \ll n_{\text{bg}} \Delta\Omega$ (i.e. detector noise negligible) \Rightarrow

the detection is background-photon-noise limited

In summary:

a) signal-photon-noise limited observation: $n_{s\min} \sim (\bar{A}_{\text{eff}} T_{\text{obs}})^{-1}$

b) background-photon-noise limited obs. : $n_{s\min} \sim (\bar{A}_{\text{eff}} T_{\text{obs}})^{-1/2}$

Note: bandwidth ΔE is implicitly incorporated $n_s = \int_{\Delta E} n_s(E) dE = \check{n}_s \Delta E$

Similar expressions for n_{bg} and n_{det} :

signal-photon-noise limited: $\check{n}_{s\min} \sim (\bar{A}_{\text{eff}} T_{\text{obs}} \Delta E)^{-1}$

background-photon-noise limited: $\check{n}_{s\min} \sim (\bar{A}_{\text{eff}} T_{\text{obs}} \Delta E)^{-1/2}$